SUPPLEMENT 2

The Dynamic Responses

Impulse Response, Step Response and Gain

Consider a simple dynamic model of the form

$$y(t) = \phi y(t-1) + x(t)\beta + \varepsilon(t). \tag{1}$$

With the use of the lag operator, we can rewrite this as

$$(1 - \phi L)y(t) = \beta x(t) + \varepsilon(t) \tag{2}$$

or, equivalently, as

$$y(t) = \frac{\beta}{1 - \phi L}x(t) + \frac{1}{1 - \phi L}\varepsilon(t). \tag{3}$$

The latter is the so-called rational transfer-function form of the equation. We can replace the operator L within the transfer functions or filters associated with the signal sequence $x(t)$ and disturbance sequence $\varepsilon(t)$ by a complex number z. Then, for the transfer function associated with the signal, we get

$$\frac{\beta}{1 - \phi z} = \beta\big\{1 + \phi z + \phi^2 z^2 + \cdots\big\}, \tag{4}$$

where the RHS comes from a familiar power-series expansion.

The sequence $\{\beta, \beta\phi, \beta\phi^2, \ldots\}$ of the coefficients of the expansion constitutes the impulse response of the transfer function. That is to to say, if we imagine that, on the input side, the signal is a unit-impulse sequence of the form

$$x(t) = \{\ldots, 0, 1, 0, 0, \ldots\}, \tag{5}$$

which has zero values at all but one instant, then its mapping through the transfer function would result in an output sequence of

$$r(t) = \{\ldots, 0, \beta, \beta\phi, \beta\phi^2, \ldots\}. \tag{6}$$

Another important concept is the step response of the filter. We may imagine that the input sequence is zero-valued up to a point in time when it assumes a constant unit value:

$$x(t) = \{\ldots, 0, 1, 1, 1, \ldots\}. \tag{7}$$

The mapping of this sequence through the transfer function would result in an output sequence of

$$s(t) = \{\ldots, 0, \beta, \beta + \beta\phi, \beta + \beta\phi + \beta\phi^2, \ldots\} \tag{8}$$

whose elements, from the point when the step occurs in $x(t)$, are simply the partial sums of the impulse-response sequence.

This sequence of partial sums $\{\beta, \beta+\beta\phi, \beta+\beta\phi+\beta\phi^2, \ldots\}$ is described as the step response. Given that $|\phi| < 1$, the step response converges to a value

$$\gamma = \frac{\beta}{1-\phi} \tag{9}$$

which is described as the steady-state gain or the long-term multiplier of the transfer function.

These various concepts apply to models of any order. Consider the equation

$$\alpha(L)y(t) = \beta(L)x(t) + \varepsilon(t), \tag{10}$$

where

$$\begin{aligned} \alpha(L) &= 1 + \alpha_1 L + \cdots + \alpha_p L^p \\ &= 1 - \phi_1 L - \cdots - \phi_p L^p, \\ \beta(L) &= 1 + \beta_1 L + \cdots + \beta_k L^k \end{aligned} \tag{11}$$

are polynomials of the lag operator. The transfer-function form of the model is simply

$$y(t) = \frac{\beta(L)}{\alpha(L)}x(t) + \frac{1}{\alpha(L)}\varepsilon(t), \tag{12}$$

The rational function associated with $x(t)$ has a series expansion

$$\begin{aligned} \frac{\beta(z)}{\alpha(z)} &= \omega(z) \\ &= \left\{\omega_0 + \omega_1 z + \omega_2 z^2 + \cdots\right\}; \end{aligned} \tag{13}$$

and the sequence of the coefficients of this expansion constitutes the impulse-response function. The partial sums of the coefficients constitute the step-response function. The gain of the transfer function is defined by

$$\gamma = \frac{\beta(1)}{\alpha(1)} = \frac{\beta_0 + \beta_1 + \cdots + \beta_k}{1 + \alpha_1 + \cdots + \alpha_p}. \tag{14}$$

The method of finding the coefficients of the series expansion of the transfer function in the general case can be illustrated by the second-order case:

$$\frac{\beta_0 + \beta_1 z}{1 - \phi_1 z - \phi_2 z^2} = \big\{\omega_0 + \omega_1 z + \omega_2 z^2 + \cdots\big\}. \tag{15}$$

We rewrite this equation as

$$\beta_0 + \beta_1 z = \big\{1 - \phi_1 z - \phi_2 z^2\big\}\big\{\omega_0 + \omega_1 z + \omega_2 z^2 + \cdots\big\}. \tag{16}$$

Then, by performing the multiplication on the RHS, and by equating the coefficients of the same powers of z on the two sides of the equation, we find that

$$\begin{aligned}
\beta_0 &= \omega_0, & \omega_0 &= \beta_0, \\
\beta_1 &= \omega_1 - \phi_1\omega_0, & \omega_1 &= \beta_1 + \phi_1\omega_0, \\
0 &= \omega_2 - \phi_1\omega_1 - \phi_2\omega_0, & \omega_2 &= \phi_1\omega_1 + \phi_2\omega_0, \\
&\vdots & &\vdots \\
0 &= \omega_n - \phi_1\omega_{n-1} - \phi_2\omega_{n-2}, & \omega_n &= \phi_1\omega_{n-1} + \phi_2\omega_{n-2}.
\end{aligned} \tag{17}$$